
APPENDIX A

The Fourier integral theorem

In equation (1.7) of Section 1.3 we gave a description of a signal defined on an infinite range in the form of a double integral, with no explanation as to how that result was obtained. Here we offer an outline derivation, taking as starting point the Fourier series representation of a function defined on a *finite* range $[-l, l]$.

From (1.6) we have

$$x(t) = \frac{1}{2l} \sum_{n=-\infty}^{\infty} \left\{ \int_{-l}^l x(t') e^{-jn\pi t'/l} dt' \right\} e^{jn\pi t/l}$$

in which n is a discrete integer variable. (It is assumed that $x(t)$ satisfies the usual conditions of continuity and differentiability except, possibly, at a limited number of points.)

It is necessary that we be able to represent functions $x(t)$ defined for *all* values of t and which are not necessarily periodic. We can proceed (in a non-rigorous manner) from the expression above as follows. On putting $u = \pi/l$, $\Delta(nu) = u$ is the frequency spacing between consecutive harmonics in the Fourier series. Since n is integer the variable (nu) is discrete, and the previous equation can be written

$$x(t) = \frac{1}{2\pi} \sum_{nu=-\infty}^{\infty} \left\{ \int_{-\pi/nu}^{\pi/nu} x(t') e^{-jn\pi t'/l} dt' \right\} e^{jn\pi t/l} \Delta(nu)$$

Now consider what happens if $l \rightarrow \infty$. We may argue, first, that the interval $(-l, l)$, which implied a period $T = 2l$, becomes $(-\infty, \infty)$ and $u \rightarrow 0$, and second, that the discrete variable nu , whose values extend over a doubly infinite range, can be replaced by a continuous variable, say $nu \rightarrow w$, so that $\Delta(nu) \rightarrow dw$ and the range of w is $(-\infty, \infty)$. Using the Riemann definition of an integral as the limit

of a sum, we then have

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \left\{ \int_{-\infty}^{\infty} x(t')^{-j\omega t'} dt' \right\} e^{j\omega t} d\omega$$

The factor $1/2\pi$ can be removed by substituting $\omega = 2\pi f$, $d\omega = 2\pi df$. The limit process is valid if $x(t)$ *does* have a Fourier series and if $\int_{-\infty}^{\infty} |x(t')| dt'$ exists.

Either in the above form (which is equation (1.7)) or rewritten in terms of f (equation (1.8)), the Fourier integral theorem is the fundamental theorem underlying all integral transform pairs (and their discrete equivalents). The various transform pairs so validated and discussed in this text are the more significant examples of what is available.

APPENDIX B

The Hartley transform

As long as it is confirmed that a proposed integral (or discrete) transform pair leads to correct recovery of the time-domain signal on inversion, one is at liberty to design transforms with either particular applications or computational implications in mind, or both. For example, once fast algorithms were developed for processing the DFT (as described in Chapters 6 and 7) there were prompted not only considerations of improved FFT algorithms but also of modifications and alternatives to the underlying (continuous) transform pair.

A well-known example is the Hartley transform, which we here describe in both continuous and discrete forms. Related to the Fourier transform (as expected), a principal feature is that ‘real’ expressions are predominant. (In fast computation of the discrete form, there is some reduction in the number of multiplications required – albeit at the expense of increased additions.)

We begin by defining a continuous transform

$$H(f) = \int_{-\infty}^{\infty} x(t') [\cos(2\pi ft') + \sin(2\pi ft')] dt' \quad (\text{B.1})$$

which is the sum of a two-sided Fourier cosine transform and a two-sided Fourier sine transform. By writing $x(t')$ as the sum of its even and odd components,

$$x(t') = \frac{1}{2}[x(t') + x(-t')] + \frac{1}{2}[x(t') - x(-t')]$$

we could write (B.1) in terms of one-sided transforms and make use of the results of Sections 1.4 and 1.6 to establish that the inverse is

$$x(t) = \int_{-\infty}^{\infty} H(f) [\cos(2\pi ft) + \sin(2\pi ft)] df \quad (\text{B.2})$$

An alternative is to show directly that these two equations satisfy the Fourier integral theorem. It was shown that this can be written in the form

$$x(t) = \frac{1}{\pi} \int_0^{\infty} \left\{ \int_{-\infty}^{\infty} x(t') \cos [w(t-t')] dt' \right\} dw$$

which appeared previously as equation (1.9). Putting $w = 2\pi f$, $dw = 2\pi df$ and noting that $\cos [2\pi f(t-t')]$ is an even function of f , this could be re-expressed as

$$x(t) = \int_{-\infty}^{\infty} \left\{ \int_{-\infty}^{\infty} x(t') \cos [2\pi f(t-t')] dt' \right\} df \quad (\text{B.3})$$

Now suppose that $H(f)$, as defined in (B.1), is substituted into the right-hand side of (B.2), and consider the expression

$$\begin{aligned} EX &= \lim_{l \rightarrow \infty} \int_{-l}^l \left\{ \int_{-l}^l x(t') [\cos (2\pi f t') + \sin (2\pi f t')] dt' \right\} \\ &\quad \cdot [\cos (2\pi f t) + \sin (2\pi f t)] df \\ &= \lim_{l \rightarrow \infty} \int_{-l}^l x(t') \left\{ \int_{-l}^l \cos [2\pi f(t-t')] + \sin [2\pi f(t+t')] df \right\} dt' \end{aligned}$$

on the changing the order of integration. Since $\sin [2\pi f(t+t')]$ is an odd function of f this reduces to

$$\begin{aligned} EX &= \lim_{l \rightarrow \infty} \int_{-l}^l x(t') \left\{ \int_{-l}^l \cos [2\pi f(t-t')] df \right\} dt' \\ &= \int_{-\infty}^{\infty} x(t') \left\{ \int_{-\infty}^{\infty} \cos [2\pi f(t-t')] df \right\} dt' \end{aligned}$$

The order of integration can be changed again, and reference to (B.3) shows that $EX = x(t)$. Equations (B.1) and (B.2) therefore constitute a transform pair.

If $H(f)$ is written in terms of its even and odd components

$$H(f) = \frac{1}{2}[H(f) + H(-f)] + \frac{1}{2}[H(f) - H(-f)] = H_e(f) + H_o(f),$$

then from equation (B.1) we see that

$$H_e(f) = \int_{-\infty}^{\infty} x(t') \cos (2\pi f t') dt'$$

and

$$H_o(f) = \int_{-\infty}^{\infty} x(t') \sin(2\pi f t') dt' \quad (\text{B.4})$$

The Fourier transform can be obtained from the equation

$$X(f) = H_e(f) - jH_o(f) \quad (\text{B.5})$$

because $e^{-j\theta} = \cos \theta - j \sin \theta$. This result holds, irrespective of whether $x(t)$ is real or complex.

If $x(t)$ is real, then $H(f)$ is real and also

$$H(f) = \text{Re}\{X(f)\} - \text{Im}\{X(f)\} \quad (\text{B.6})$$

Equation (B.6) does *not* apply if $x(t)$ is a complex signal.

The Hartley transform is more symmetrical than the Fourier transform because the transform pair, (B.1) and (B.2), have an identical structure.

The discrete form of the Hartley transform is defined by the equation

$$H(n) = \sum_{k=0}^{N-1} x_k \left[\cos\left(\frac{2\pi nk}{N}\right) + \sin\left(\frac{2\pi nk}{N}\right) \right]$$

or, with an obvious notation,

$$H(n) = \sum_{k=0}^{N-1} x_k \text{cas}\left(\frac{2\pi nk}{N}\right) \quad (\text{B.7})$$

and the inversion is provided by

$$x(k) = \frac{1}{N} \sum_{n=0}^{N-1} H(n) \text{cas}\left(\frac{2\pi nk}{N}\right) \quad (\text{B.8})$$

(In some texts, the multiplier $1/N$ appears in the definition of $H(n)$. Here it has been transferred to the inversion-sum, (B.8), to facilitate comparison with the DFT as defined in (5.5).)

These equations again show complete symmetry, and no modification (such as conjugation) is needed to use any particular algorithm for both transformation and inversion. To verify that we have a transform pair we begin as we did (in the case of the DFT) in Section 5.2. In (B.7) we replace the summation integer k by m and substitute into the right-hand side of (B.8), in which k is some fixed integer, to obtain

$$EX = \frac{1}{N} \sum_{n=0}^{N-1} \left\{ \sum_{m=0}^{N-1} x_m \text{cas}\left(\frac{2\pi nm}{N}\right) \right\} \text{cas}\left(\frac{2\pi nk}{N}\right)$$

If the order of summation is changed,

$$EX = \frac{1}{N} \sum_{m=0}^{N-1} x_m \left\{ \sum_{n=0}^{N-1} \operatorname{cas}\left(\frac{2\pi nm}{N}\right) \operatorname{cas}\left(\frac{2\pi nk}{N}\right) \right\} \quad (\text{B.9})$$

The summation over n can be written

$$\sum(m) = \sum_{n=0}^{N-1} \left\{ \cos\left[\frac{2\pi n(k-m)}{N}\right] + \sin\left[\frac{2\pi n(k+m)}{N}\right] \right\}$$

if the terms are expanded and simplified using trigonometric formulae. Considering the first term in $\sum(m)$, we can write

$$\sum_{n=0}^{N-1} \cos\left[\frac{2\pi n(k-m)}{N}\right] = \operatorname{Re} \left\{ \sum_{n=0}^{N-1} e^{j2\pi n(k-m)/N} \right\}$$

This sum appeared in (5.7) and, by considering the sum of a geometric progression, was shown to be zero if $k \neq m$ and N if $k = m$. Similarly, the second term in $\sum(m)$ can be written

$$\sum_{n=0}^{N-1} \sin\left[\frac{2\pi n(k+m)}{N}\right] = \operatorname{Im} \left\{ \sum_{n=0}^{N-1} e^{j2\pi n(k+m)/N} \right\}$$

in which the exponentials form another geometric progression, whose sum is

$$S_N = \frac{1 - e^{j2\pi(k+m)}}{1 - e^{j2\pi(k+m)/N}}$$

(unless this is an indeterminate ratio). The numerator is always zero. The denominator is zero only if $(k+m) = N$, and k and m can assume values in the range $0, 1, 2, \dots, N-1$ only. In that case,

$$\sum_{n=0}^{N-1} e^{j2\pi n(k+m)/N} = \sum_{n=0}^{N-1} e^{j2\pi n} = \sum_{n=0}^{N-1} (-1)^n$$

which is zero as N is an even number. Hence $\sum(m)$ is zero if $m \neq k$ and has value N if $m = k$. Equation (B.9) reduces to $EX = x_k$, and that (B.7) and (B.8) are a transform pair is confirmed.

(The orthogonality relation,

$$\sum_{n=0}^{N-1} \operatorname{cas}\left(\frac{2\pi nm}{N}\right) \operatorname{cas}\left(\frac{2\pi nk}{N}\right) = N\delta_{mk}$$

is analogous to the orthogonality relations used when finding expressions for the coefficients $\{a_n\}$ and $\{b_n\}$ in a trigonometric Fourier series.)

APPENDIX C

Further reading

In the preface it was explained that there is an extensive literature on the subject of discrete signals, systems and transforms but that the emphasis on particular applications and the level of assumed knowledge (mathematical or otherwise) are highly variable. What follows is a short selection of titles intended to reflect this diversity.

- Bateman, A. and Yates, W. (1988) *Digital Signal Processing Design*, Pitman.
- Bracewell, R.N. (1986) *The Fourier Transform and its Applications* (2nd edn), McGraw-Hill.
- Brigham, E.O. (1974) *The Fast Fourier Transform*, Prentice Hall.
- Doetsch, G. (1961) *Guide to the Applications of Laplace Transforms*, Van Nostrand.
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- Proakis, J.G. and Manolakis, D.G. (1988) *Introduction to Digital Signal Processing*, Macmillan.
- Roberts, R.A. and Mullis, C.T. (1987) *Digital Signal Processing*, Addison Wesley.
- Soliman, S. and Srinath, M.D. (1990) *Continuous and Discrete Signals and Systems*, Prentice Hall.
- Strum, R. and Kirk, D. (1988) *First Principles of Discrete Systems and Digital Signal Processing*, Addison Wesley.
- Ziemer, R.W., Tranter, W.H. and Fannin, D.R. (1983) *Signals and Systems, Continuous and Discrete*, Macmillan.

Index

- Aliasing 60–62, 66
- Amplitude spectrum 2, 57–67
- Analogue/digital conversion, *see*
 - Sampling
- Autocorrelation 132

- Bandlimited signals, *see* Signal
- Binary code 56
- Bit reversal 56
- Butterfly 151–3

- Convergence
 - Laplace transform 14
 - Fourier transform 32–5
- Convolution
 - continuous 13, 31, 44–51
 - discrete 82–5
 - graphical 44–51, 83–4
- Correlation 132
- Cosine transform, *see* Fourier cosine transform
- Cross-correlation 132

- Decimation 142, 157–8
 - in frequency 142–50, 169–75
 - in time 142, 151, 158–69
- Delta function, *see* Function
- Difference equations 91–112
 - ad hoc* solution 94–6
 - characteristic polynomial 97
 - Z-transform solution 96–112

- Differential equations
 - approximation by difference equations 92–3
 - solution by Laplace transform 16–19
- Discrete Fourier transform 117–37
 - inversion 120–21, 126–7
 - matrix factorization 141–51
 - properties 129–32
- Discrete Hartley transform 182–3
- Discrete signals, *see* Sampling

- Encoding 56
- Energy
 - continuous signal 67
 - discrete signal 133

- Fast Fourier transform 156–75
 - decimation in frequency 169–75
 - decimation in time 158–69
 - matrix factorization 141–51
 - signal flow graph 162–75
- Flow graph, *see* Signal flow graph
- Fourier cosine transform 11
- Fourier integral theorem 6–8, 178–9
- Fourier integral transform 9
 - convolution 31
 - inversion 9
 - pairs 36–44
 - properties 26–32

- Fourier Series
 - exponential 3, 118
 - trigonometric 2
 - Gibbs' phenomenon 68
- Fourier sine transform 11
- Fourier transform, *see* Fourier integral transform
- Function
 - delta 15
 - even 8
 - odd 8
 - rectangle 38
 - signum 34
 - sinc 40
 - sine-integral 41
 - step 13
 - transfer, *see* System, transfer function
 - triangle 39
 - window 65, 69, 135
- Graph, *see* Signal flow graph
- Graphical convolution, *see* Convolution
- Gibbs' phenomenon 68
- Hartley transform
 - continuous 180–82
 - discrete 182–3
- Ideal sampling, *see* Sampling
- Impulse function, *see* Function, delta
- Impulse response 70
- Impulse train 57
- Integral
 - convolution 13, 31
- Inverse transform
 - discrete Fourier 120–21, 126–7, 168, 175
 - Fourier 9, 68
 - Laplace 12
 - Z-transform 79–81
- Laplace transform 11
 - convolution theorem 13, 20
 - inversion 12
- pairs 15
- properties 13–16
- Magnitude, *see* Amplitude spectrum
- Matrix
 - coefficients of the DFT 143
 - factorization 141–51
 - permutation 125, 144
- Nyquist sampling rate, *see* Sampling
- Operator
 - backward difference 92
 - forward difference 92
- Parseval's theorem 67, 133–4
- Periodic functions
 - Fourier series 2
 - transform of ideally sampled signal 59
- Permutation matrix 125, 144
- Phase spectrum 3
- Power 133
- Pulse train 57
- Quantization 55–6
- Rectangle function, *see* Function
- Recursion 141
- Response, *see* System, response
- Sampling
 - aliasing, *see* Aliasing
 - frequency 42, 55
 - functions 56–9, 134–5
 - ideal 42, 59
 - interval 42, 55
 - Nyquist rate 60
 - reconstruction theorem 71
- Series
 - binomial 73, 74
 - Fourier, *see* Fourier series
 - geometric 73, 120
- Signal
 - band limited 58
 - energy of 67, 133

- flow graph, *see* Signal flow graph
- representation by integral 6–8
- representation by series 2
- sampled, *see* Sampling
- Signal flow graph 157
 - butterfly diagram 151–3
 - decimation in frequency 169–75
 - decimation in time 162–8
 - inversion processes 168–9
 - twiddle factors 153
- Sine transform, *see* Fourier sine transform
- Spectral leakage 66
- Spectrum, *see* Amplitude spectrum, Phase spectrum
- System
 - linear 17
 - response 74
 - transfer function 19, 74–5
- Tables
 - discrete Fourier transform
 - properties 129
 - Fourier integral transform pairs 41
 - Fourier integral transform
 - properties 27
 - Laplace transform pairs 15
 - Laplace transform properties 13
 - Z-transform pairs 78
 - Z-transform properties 76
- Transfer function, *see* System
- Transform, *see* Discrete Fourier transform, Fast Fourier transform, Fourier cosine transform, Fourier integral transform, Fourier sine transform, Hartley transform, Laplace transform, Z-transform
- Truncation
 - Fourier series 118–19
 - integration sum 63
 - windowing 65–70, 134–5
- Twiddle factor 153
- Window functions 65, 69, 135
- Z-transform 72–85
 - convolution 82–5
 - inversion 79–81
 - pairs 73–4, 78–9
 - properties 75–7